

## FEATURES

- Excellent noise performance through the entire gain range
- Exceptionally low THD+N over the full audio bandwidth
- Low power consumption – 6 mA typical
- Wide bandwidth – 3 MHz typ. @ G=1000
- High Slew Rate – 28 V/μs @ G=10 and C<sub>L</sub>=50pF
- Wide Output Swing – ±13 V typ. on ±15 V supplies
- Gain adjustable from 1 to >1000 with a single external resistor
- Drop-in compatible with SSM-2017

## APPLICATIONS

- Mixing Consoles
- P/A Systems
- Analog and Digital Snakes
- Breakout Boxes for Digital Systems
- Instrumentation

## Description

The THAT 1510 is a high-performance audio preamplifier suitable for microphone preamp and bus summing applications. The IC is pin compatible with the Analog Devices SSM2017 in both the 8-pin DIP and 16-pin SOIC packages. An 8-pin SOIC package is also available, providing a compact form factor for new designs.

Designed from the ground up in THAT's complementary bipolar, dielectric isolation process, the THAT 1510 improves on existing integrated

microphone preamps by offering lower noise at low gains, better distortion characteristics, lower power consumption, higher slew rate and bandwidth, and increased output voltage swing. The part is fully protected against ESD on all critical pins, and reliability is further enhanced by its reduced power requirements.

In short, the THAT 1510 provides superior performance in a popular format at an affordable price.

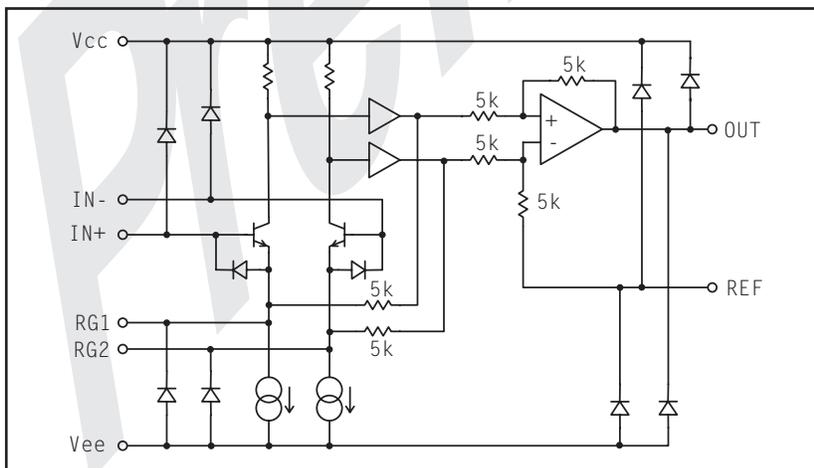


Figure 1. THAT 1510 Equivalent Circuit Diagram

Pin Name	DIP	SO8 Pin	SO16 Pin
RG <sub>1</sub>	1	1	2
-In	2	2	4
+In	3	3	5
V-	4	4	7
Ref	5	5	10
Out	6	6	11
V+	7	7	13
RG <sub>2</sub>	8	8	15

Table 1. THAT 1510 pin assignments

**SPECIFICATIONS<sup>1</sup>****Absolute Maximum Ratings ( $T_A = 25^\circ\text{C}$ )**

Positive Supply Voltage ( $V_{CC}$ )	+20 V	Operating Temperature Range ( $T_{OP}$ )	-40 to +85°C
Negative Supply Voltage ( $V_{EE}$ )	-20 V	Storage Temperature Range ( $T_{ST}$ )	-40 to +125°C
Output Short-Circuit Duration ( $t_{SH}$ )	Continuous	Junction Temperature ( $T_J$ )	150°C
Power Dissipation ( $P_D$ )	TBD mW	Lead Temp. ( $T_{LEAD}$ ) (Soldering 60 sec)	TBD °C
Max Input Swing (IN+ and IN-)	$V_{CC} + .3V$ to $V_{EE} - .3V$		

**Recommended Operating Conditions**

Parameter	Symbol	Conditions	Min	Typ	Max	Units
Positive Supply Voltage	$V_{CC}$		+5		+18	V
Negative Supply Voltage	$V_{EE}$		-5		-18	V

**Electrical Characteristics<sup>2</sup>**

Parameter	Symbol	Conditions	Min	Typ	Max	Units
Supply Current	$I_{CC}$	No signal	—	6.1	8.2	mA
	$I_{EE}$		—	—	-8.4	mA
Input Bias Current	$I_B$	No signal; Either input connected to GND	—	8.6	25	$\mu\text{A}$
Input Offset Current	$I_{OS}$	No signal	—	—	$\pm 2.5$	$\mu\text{A}$
Output Referred Offset Voltage	$V_{OS_{OR}}$	No Signal, $V_{CM}=0$	—	—	$\pm 50$	mV
Input Referred Offset Voltage	$V_{OS_{IR}}$		—	—	$\pm 1.2$	mV
Input Voltage Range Common Mode	$V_{IN-CM}$	Common mode, all gains	—	$\pm 13$	—	V
	$V_{IN-UNBAL}$	Unbalanced One input to GND, 0dB gain	—	$\pm 13$	—	V
Differential Gain	$G_{diff}$		0	—	70	dB
Ref Input Voltage Range			—	$\pm 8$	—	V
Ref Input Impedance			—	10	—	k $\Omega$
Ref Input Gain to Output			—	0	—	dB
Input Impedance	$Z_{IN-DIFF}$	Differential 0dB gain	—	32  1.9	—	M $\Omega$   pF
		20dB gain	—	32  2.0	—	M $\Omega$   pF
		40dB gain	—	32  2.5	—	M $\Omega$   pF
		60dB gain	—	29  8.0	—	M $\Omega$   pF
	$Z_{IN-CM}$	Common mode all gains	—	8  7.7	—	M $\Omega$   pF

- All specifications are subject to change without notice.
- Unless otherwise noted,  $T_A=25^\circ\text{C}$ ,  $V_{CC} = +15V$ ,  $V_{EE} = -15V$

<b>Electrical Characteristics (Cont'd)</b>							
Parameter	Symbol	Conditions	Min	Typ	Max	Units	
Common Mode Rejection	CMR	$V_{CM} = \pm 10V$ ; DC to 60 Hz	0 dB gain	20	55	—	dB
			20 dB gain	40	75	—	dB
			40 dB gain	60	95	—	dB
			60 dB gain	80	115	—	dB
Power Supply Rejection	PSR	$V_{CC} = -V_{EE} = \pm 5V$ to $\pm 20V$ ; DC to 60 Hz	0 dB gain	—	85	—	dB
			20 dB gain	—	105	—	dB
			40 dB gain	—	120	—	dB
			60 dB gain	—	124	—	dB
Total Harmonic Distortion	THD	$V_{OUT} = 7V_{rms}$ ; $R_L = 5 k\Omega$ $f = 1kHz$ ; BW = 20 kHz	0 dB gain	—	0.0005	—	%
			20 dB gain	—	0.0005	—	%
			40 dB gain	—	0.0005	—	%
			60 dB gain	—	0.005	—	%
		$f = 20kHz$ ; BW = 80 kHz	0 dB gain	—	0.002	—	%
			60 dB gain	—	0.005	—	%
Equivalent Input Noise	$e_{nOUT}$	$f = 1kHz$ , 0dB gain	—	57	—	nV/ $\sqrt{Hz}$	
			20dB gain	—	7	—	nV/ $\sqrt{Hz}$
			40dB gain	—	1.7	—	nV/ $\sqrt{Hz}$
			60dB gain	—	1	—	nV/ $\sqrt{Hz}$
Input Current Noise	$i_n$	60dB gain	—	2.3	—	pA/ $\sqrt{Hz}$	
Noise Figure	NF	60dB gain	—	1.6	—	dB	
		$R_S = 150 \Omega$	—	1.3	—	dB	
		$R_S = 200 \Omega$	—		—		
Slew Rate	SR	$R_L = 2 k\Omega$ $C_L = 50 pF$	16	28	—	V/ $\mu s$	
Bandwidth -3dB	BW-3dB	$R_L = 2 k\Omega$ ; $C_L = 10 pF$	0dB gain	8	15	—	MHz
			20dB gain	6	8	—	MHz
			40dB gain	5	7	—	MHz
			60dB gain	2	3	—	MHz
Output Gain Error	$G_{EROUT}$	$f = 1kHz$ ; $R_L = 2 k\Omega$	0dB gain	—	$\pm 0.25$	$\pm 1$	dB
			20dB gain	—	$\pm 0.20$	$\pm 1$	dB
			40dB gain	—	$\pm 0.20$	$\pm 1$	dB
			60dB gain	—	$\pm 0.05$	$\pm 0.5$	dB
Output Voltage Swing	$V_O$	$R_L = 2 k\Omega$ all gains	$\pm 13$	$\pm 13.3$	—	V	
Output Short Circuit Current	$I_{SC}$	$R_L = 0 \Omega$	—	$\pm 17$	—	mA	
Minimum Resistive Load	$R_{Lmin}$		—	2	—	k $\Omega$	
Maximum Capacitive Load	$C_{Lmax}$		—	—	200	pF	
Gain Equation		$G = 1 + \frac{10k\Omega}{R_G}$					
Gain Resistor	$R_G$	0dB gain		$\infty$		$\Omega$	
		20dB gain		1,100		$\Omega$	
		40dB gain		101		$\Omega$	
		60dB gain		10		$\Omega$	

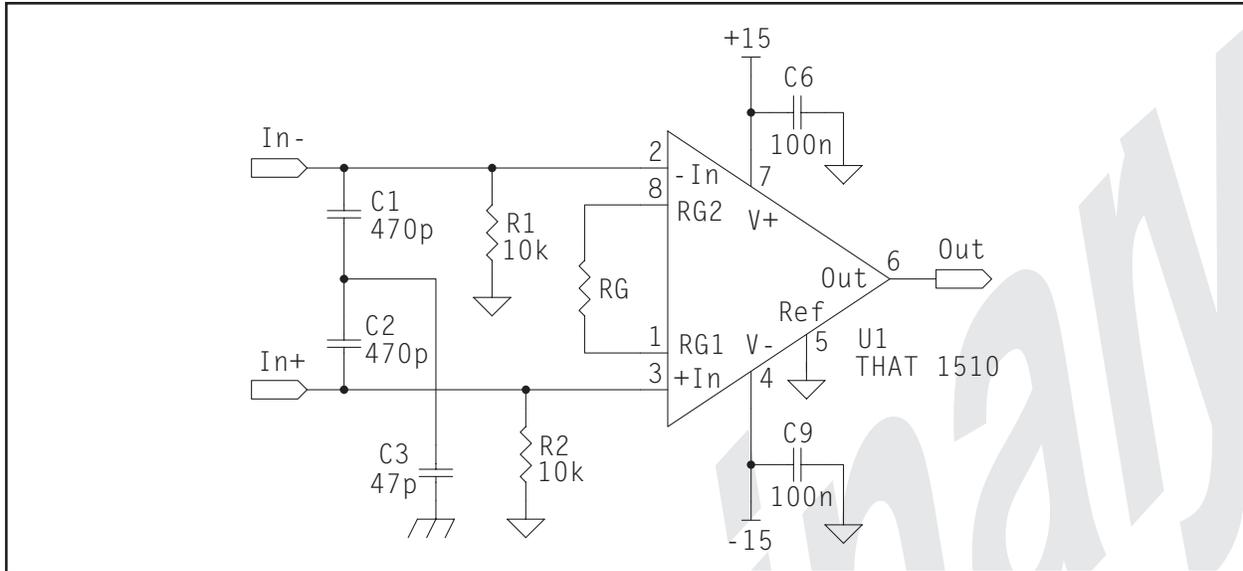


Figure 2. Basic 1510 Circuit

### Applications

#### Gain Setting

A single external resistor between the RG1 and RG2 pins is all that is needed to set the gain of the THAT 1510, according to the formula

$$G = \frac{1+10k\Omega}{RG} \quad \text{or}$$

$$RG = \frac{10k\Omega}{G-1}$$

RG1 and RG2 may be left open for unity gain operation. To avoid excess noise and ensure tempera-

ture stability, wirewound or metal-film resistors are recommended for RG.

Total gain accuracy will depend on both the tolerance on RG, and on the gain equation accuracy of the THAT 1510. Total gain drift will result from the mismatch between the tempco of RG and the tempco of the internal resistors ( $\pm 20$  ppm/ $^{\circ}$  C typical).

#### Noise Performance

The THAT 1510 has significantly lower noise at low gains than similar IC microphone preamps. At

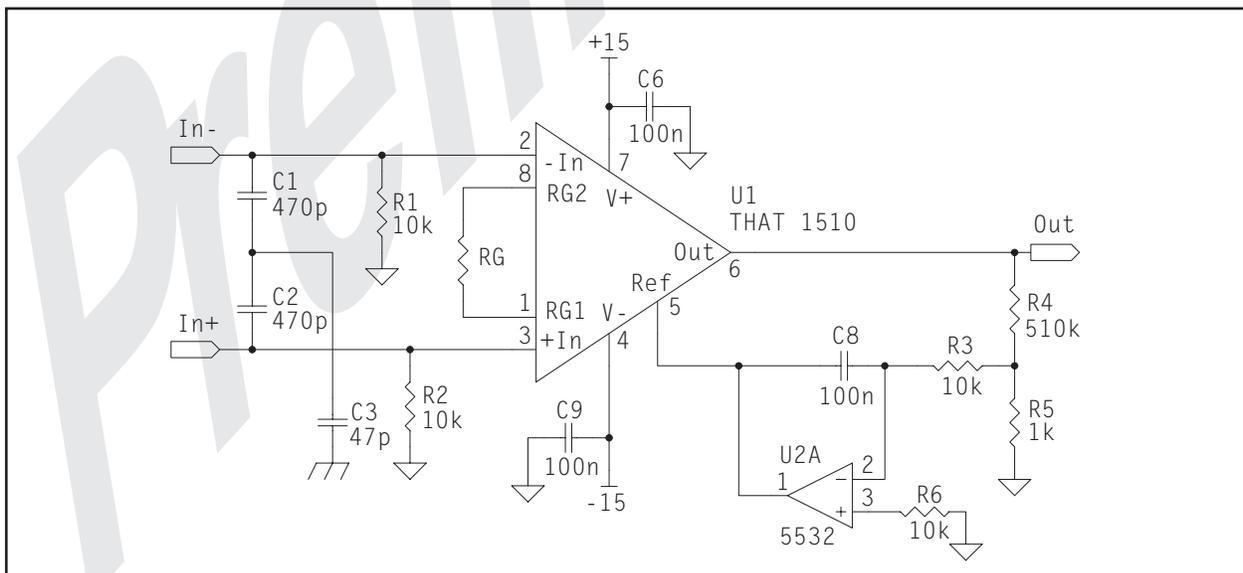


Figure 3. 1510 Circuit with Output Offset Correction

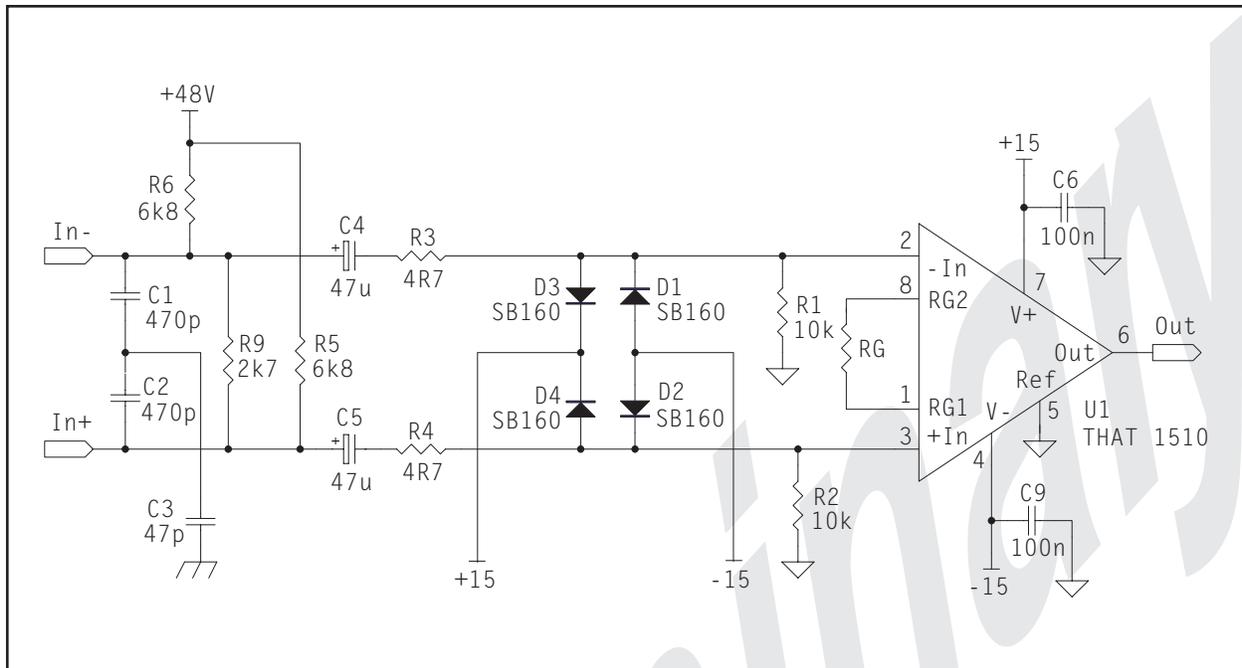


Figure 4. 1510 Circuit with Phantom Power

1 kHz, its equivalent input noise is 1 nV/√Hz at 60 dB gain. At 0 dB gain, equivalent input noise is 55 nV/√Hz, nearly 6dB better than competitive IC designs

## Inputs

Protection diodes are employed at all pins except V+ and V- of the THAT 1510. These diodes prevent accidental ESD/EOS damage to the IC. Other diodes across the base-emitter junctions of the input transistors prevent reverse breakdown of these junctions, which could degrade noise performance.

The THAT 1510's inputs are floating, so a dc bias connection is required to maintain the inputs within the IC's input common-mode range. Three different schemes are shown in Figures 2, 3, and 4. Note that the values of R1 and R2 in these figures should be kept small to limit common-mode pickup. A value of 10 kΩ is recommended.

## Reference Terminal

The "Ref" pin provides the reference for the output signal, and is normally connected to analog ground. If necessary, the "Ref" pin can be used for offset correction or DC level shifting. If this is done, the "Ref" pin should be driven from a low impedance source such as an op-amp output to avoid degrading CMRR.

## Phantom Power

Phantom power is required for condenser microphones. Figure 4 shows the recommended phantom power circuit for the THAT1510. When there is provision for phantom power, the designer must beware of the possibility of a phantom power fault. These faults usually occur when a mic pre channel is floating with phantom power applied (a common situation in portable consoles with a single phantom power switch for all mic inputs), and is accidentally shorted to a line driver output or ground. A large portion of the energy stored on the coupling capacitors can be shunted through internal junctions in the IC's, resulting in degradation and failure. Note that this energy is often more than 100 mJ, an order of magnitude greater than the energy in the most stringent (optional) ESD tests.

The SB160s, D1 through D4, comprise a bridge that shunts the potentially destructive currents that can result from a phantom power fault directly to the power supplies. If external protection is not provided, these currents will find a path through the IC, often through the on-chip ESD diodes, and sometimes through the input devices themselves. Since these currents can be as high as several amps in a low noise circuit, bond wires or on-chip metal can fail. With Schottky diodes in place, current into the input will be limited to less than 100 mA, which is well within the capability of the process.

### Package Information

The THAT1510 is available in 8-pin mini-DIP, 8-pin SOIC, and 16-pin SOIC packages. The package

dimensions are shown in Figures 5, 6, and 7 while pinouts are given in Table 1.

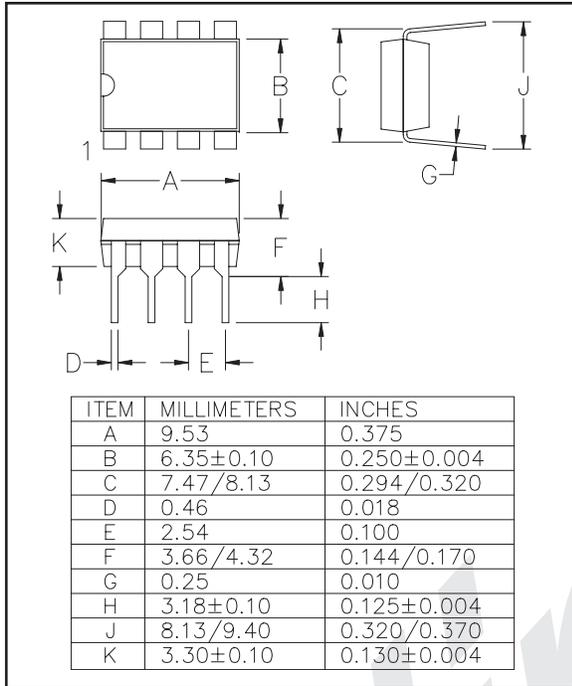


Figure 5. -P (8-pin DIP) version package outline

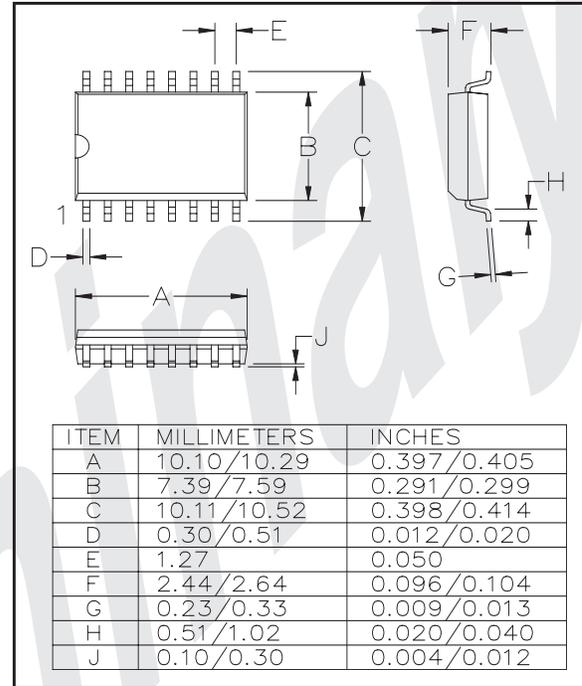


Figure 6. -S16 (16-pin SO) version package outline

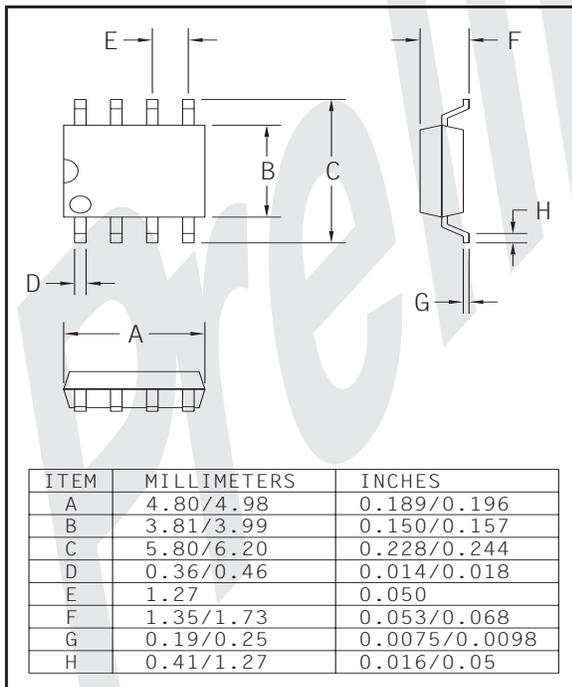


Figure 7. -S8 (8-pin SO) version package outline